

Chap. 4 Multirate Filters

Filt 33

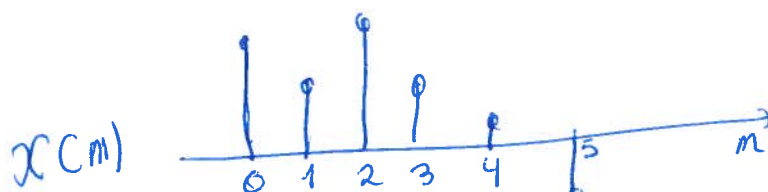
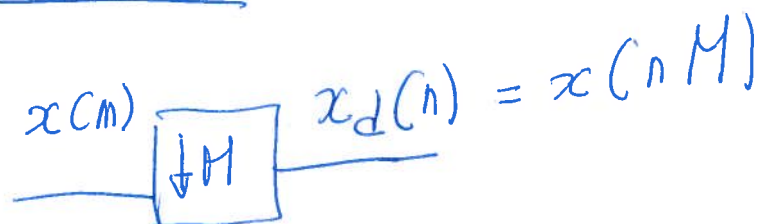
4.1

Why

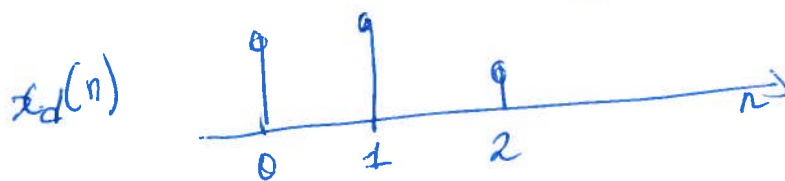
- + Different sampling rates co-exist
- + Wideband vs. narrowband \Rightarrow Multiplexing
- + Multi resolution filters
- + Efficient filters

4.2

Decimation



$M=2$



$$\Rightarrow X_d(e^{j\omega}) = \frac{1}{M} \sum_{k=0}^{M-1} X(e^{j(\omega - 2\pi k)/M})$$

$$x'(m) = \begin{cases} x(m) & m = nM, n \in \mathbb{Z} \\ 0 & \text{otherwise} \end{cases} \quad x'(m) = x(m) \sum_{l=-\infty}^{+\infty} \delta(m-lM)$$

$$\begin{aligned} 1 \quad X_d(e^{j\omega}) &= \sum_n x_d(n) e^{-j\omega n} \\ &= \sum_n x(nM) e^{-j\omega n} \\ &= \sum_n x'(nM) e^{-j\omega n} \\ &= \sum_l x'(l) e^{-j\omega \frac{l}{M}} = X'(e^{j\omega}) \end{aligned}$$

$$\begin{aligned} 2 \quad X'(e^{j\omega}) &= X(e^{j\omega}) * \text{TF} \left\{ \sum \delta(n-nM) \right\} \\ &= X(e^{j\omega}) * \frac{2\pi}{M} \sum_{k=0}^{M-1} \delta(\omega - \frac{2\pi k}{M}) \\ &= \frac{1}{M} \sum_{k=0}^{M-1} X(e^{j(\omega - \frac{2\pi k}{M})}) \end{aligned}$$

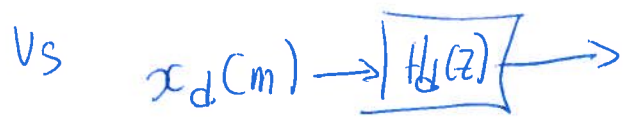
$$(1) (2) \Rightarrow X_d(e^{j\omega}) = \frac{1}{M} \sum_{k=0}^{M-1} X(e^{j(\omega - \frac{2\pi k}{M})})$$

! Aliasing

Graph on Blackboard

⇒ "widening" of spectrum in ω

! Complexity → Discuss complexity of



4.2.2 On Aliasing

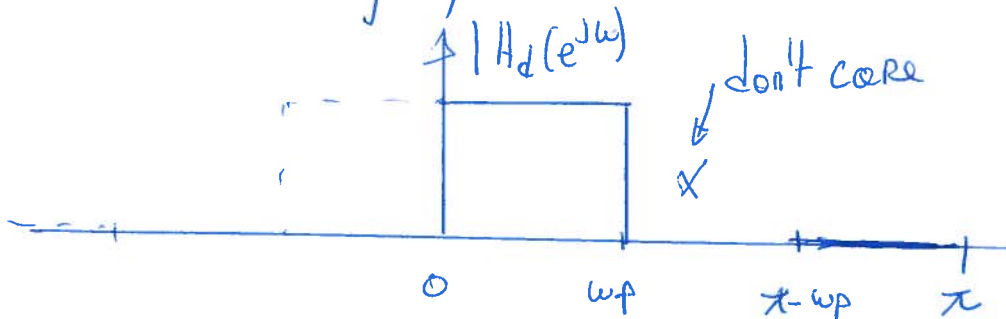


Anti-aliasing Filter $H_d(z)$?

Let $X(z) = 0$ for $|\omega| > \omega_p$, $\omega_p < \frac{\pi}{M}$

$\Rightarrow H_d(e^{j\omega})$ can be relaxed to

$$H_d(e^{j\omega}) = \begin{cases} 1, & |\omega| < \omega_p \\ 0, & |\omega| \in \left[\frac{2\pi k}{M} - \omega_p, \frac{2\pi k}{M} + \omega_p \right] \quad k=1, 2, \dots, M \end{cases}$$

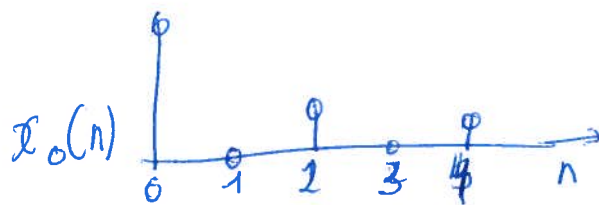
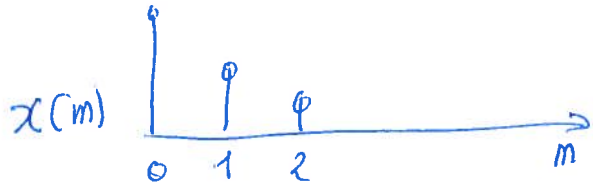
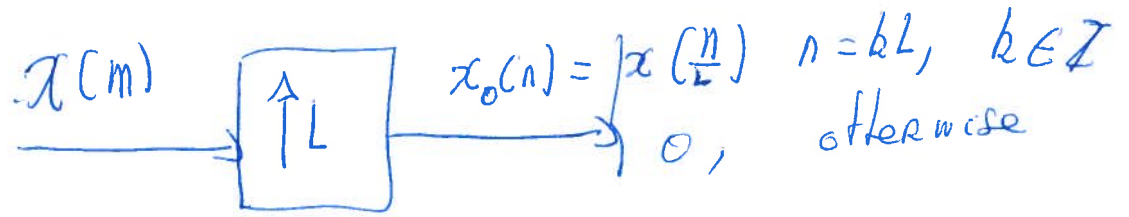


\rightarrow draw aliasing on black board

\rightarrow show slide 21-23

4.3 Over-sampling

4.3.1 Basics



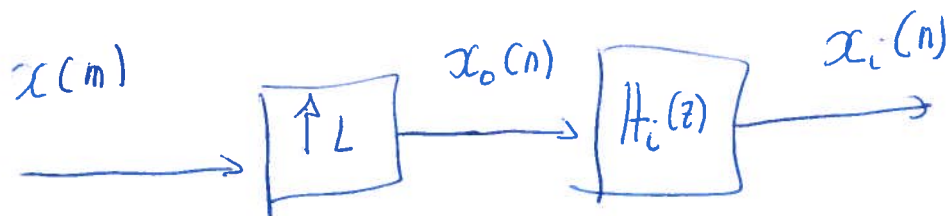
L = 2

$$X_o(e^{j\omega}) = X(e^{j\omega L})$$

straight forward,
write $X_o(z)$ vs $X(z)$

"narrowing" of spectrum in ω

4.3.2 Interpolation

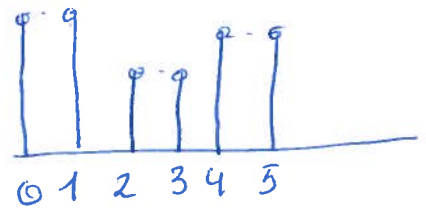


$$H_i(e^{j\omega}) = \begin{cases} L & |\omega| < \pi/L \\ 0 & \text{otherwise} \end{cases}$$

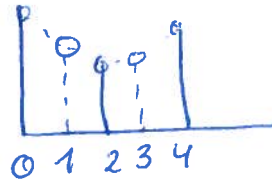
Discuss Complexity

Other interpolators

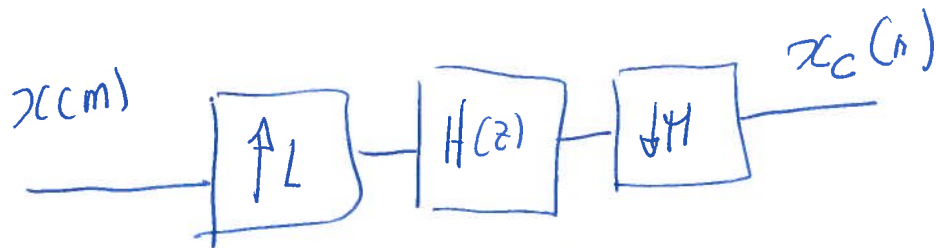
- Zero-hold



- Linear



4.4. Rational sampling-rate change



$H(z)$ } Interpolation filter $\omega_c < \frac{\pi}{L}$
 } decimation filter $\omega_c < \frac{\pi}{M}$

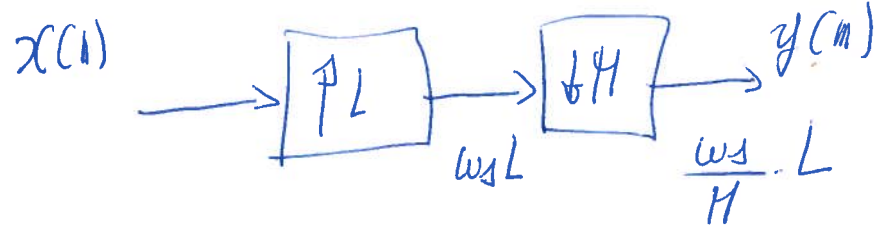
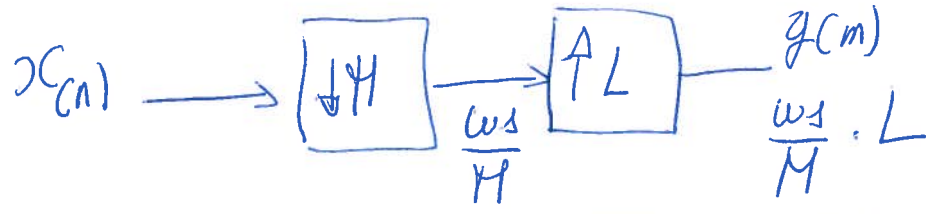
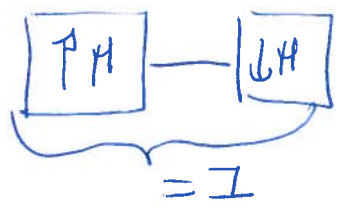
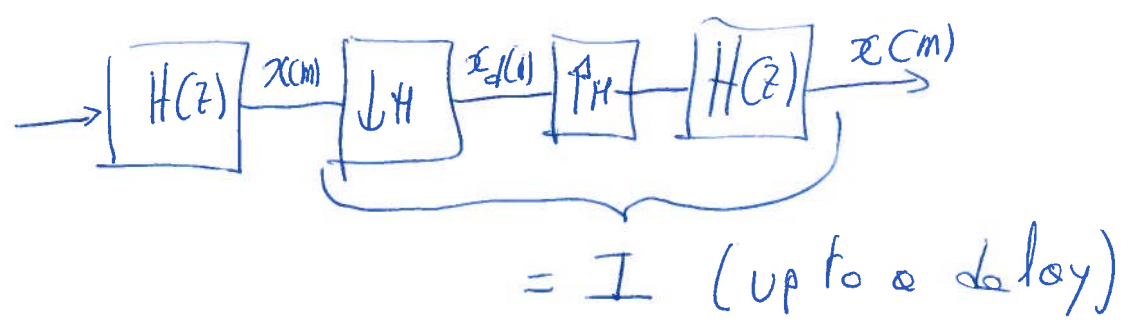
$$\Rightarrow H(z) = \begin{cases} L & |\omega| < \min\left(\frac{\pi}{L}, \frac{\pi}{M}\right) \\ 0 & \text{otherwise} \end{cases}$$

→ Relaxation Possible

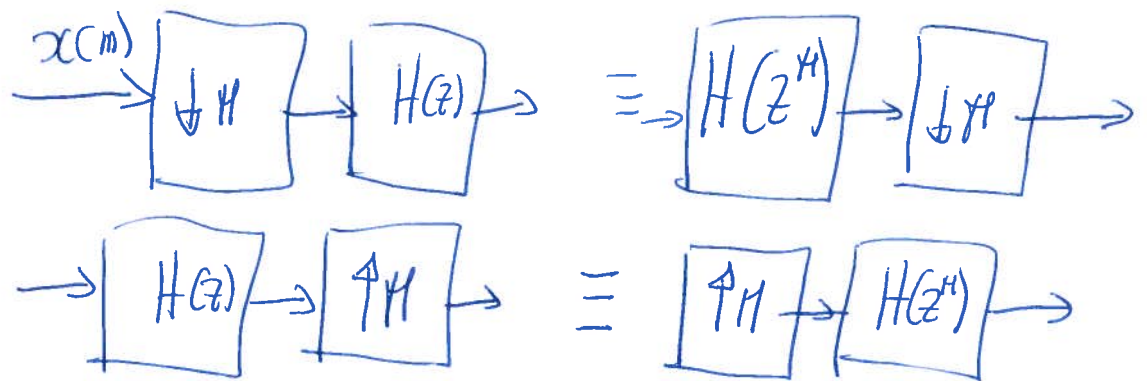
$$H(z) = \begin{cases} L & |\omega| < \min\left(\frac{\omega_p}{L}, \frac{\pi}{M}\right) \\ 0 & \text{otherwise} \end{cases}$$

$$\min\left\{\frac{2\pi}{L} - \frac{\omega_p}{L}, \frac{2\pi}{M} - \frac{\omega_p}{L}\right\} \leq |\omega| \leq \pi$$

4.5 Inverse Operations



4.6 Noble Identities



4.7 Polyphase decomposition

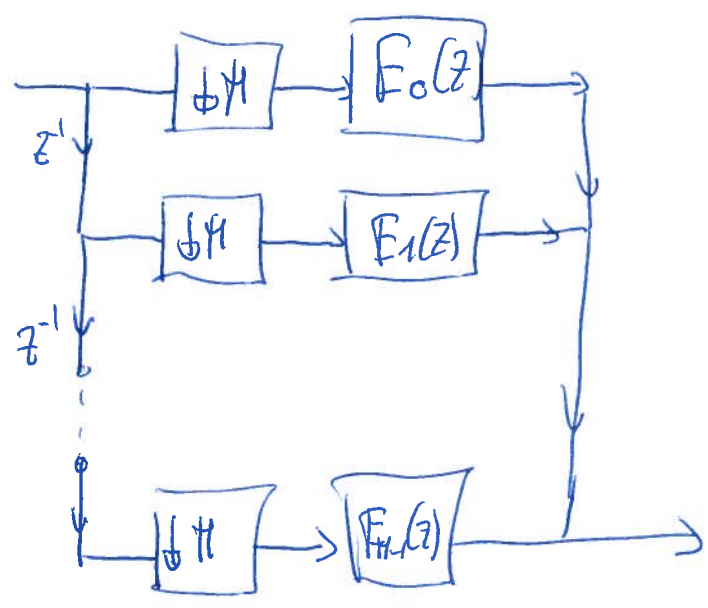
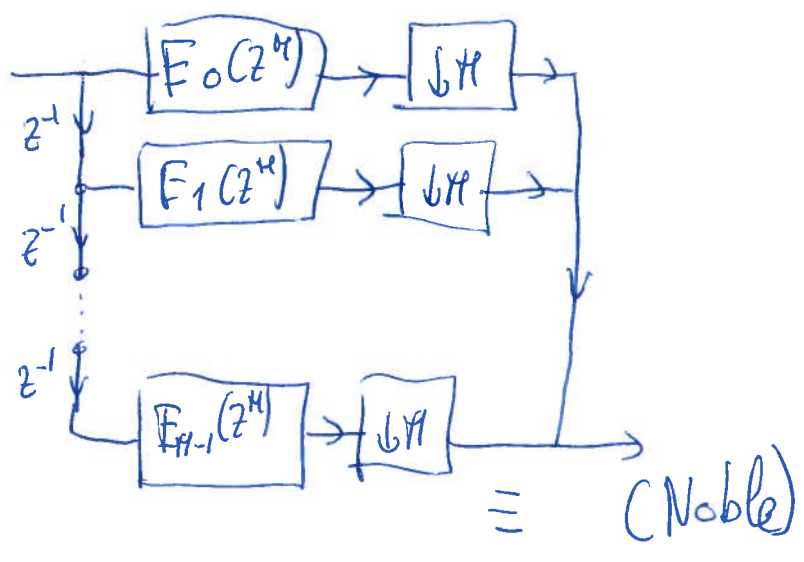
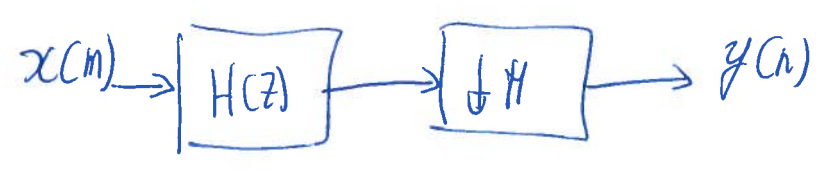
We can write $H(z)$ as

$$\begin{aligned}
 H(z) &= \sum h(k) z^{-k} \\
 &= \sum_{\ell} h(M\ell) z^{-M\ell} + \sum_{\ell} h(M\ell+1) z^{-(M\ell+1)} + \dots + \sum_{\ell} h(M\ell+M-1) z^{-(M\ell+M-1)} \\
 &= \sum_{\ell} h(M\ell) z^{-M\ell} + z^{-1} \sum_{\ell} h(M\ell+1) z^{-M\ell} + \dots + z^{-(M-1)} \sum_{\ell} h(M\ell+M-1) z^{-M\ell} \\
 &= \sum_{j=0}^{M-1} z^{-j} F_j(z^M)
 \end{aligned}$$

where $F_j(z^M) = \sum_{\ell=-\infty}^{+\infty} h(M\ell+j) z^{-\ell M}$

= Type I polyphase filter

Decimation by Polyphase dec.

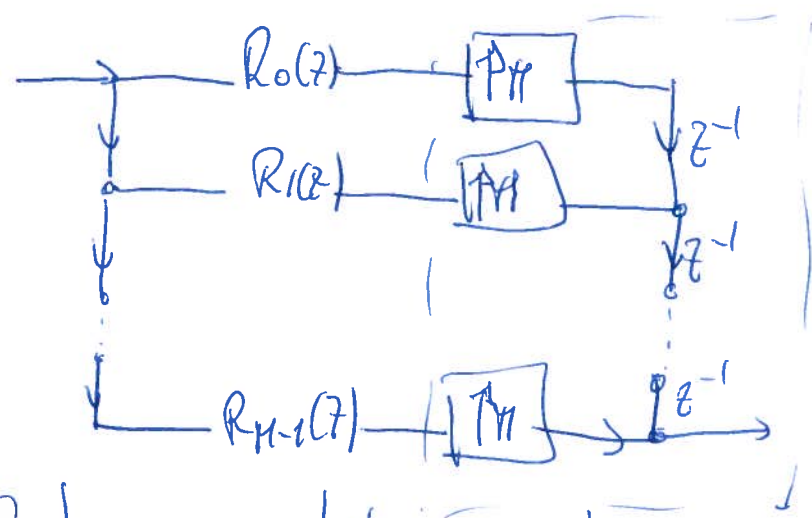


Polyphase Component of type II

$$R_j(z) = E_{M-1-j}(z)$$

$$\Rightarrow H(z) = \sum_{i=0}^{M-1} z^{-(M-1-i)} R_j(z^M)$$

Interpolation by Polyphase Dec.



Rotary switches

